Aeroservoelastic Model Uncertainty Bound Estimation from Flight Data

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Uncertainty modeling is a critical element in the estimation of robust stability margins for stability boundary prediction and robust flight control system development. There has been a serious deficiency to date in aeroservoelastic data analysis with attention to uncertainty modeling. Uncertainty can be estimated from flight data using both parametric and nonparametric identification techniques. The model validation problem addressed here is to identify aeroservoelastic models with associated uncertainty structures from a limited amount of controlled excitation inputs over an extensive flight envelope. The challenge is to update analytical models from flight data estimates while also deriving nonconservative uncertainty descriptions consistent with the flight data. Transfer function estimates are incorporated in a robust minimax estimation scheme to update models and get error bounds consistent with the data and model structure. Uncertainty estimates derived from the data in this manner provide an appropriate and relevant representation for model development and robust stability analysis. The method incorporates parametric and nonparametric uncertainty into various uncertainty structures for quantitative measures of robust stability relating to parameter variations and unmodeled dynamics. This model-plus-uncertainty identification procedure is applied to aeroservoelastic flight data from the NASA Dryden Flight Research Center F-18 Systems Research Aircraft.

Nomenclature

Nomenciature			
$\{A, B, C, D\}$	=	plant state-space matrices estimated plant state-space matrices	
$\{\hat{A},\hat{B},\hat{C},\hat{D}\}$	=	estimated plant state-space matrices	
K		controller	
P P	=	aeroservoelastic plant	
\hat{P}	=	updated aeroservoelastic plant	
T	=	complementary sensitivity function	
W	=	uncertainty weightings	
y	=	output frequency response vector	
ΔC	=	uncertainty in C matrix	
$\Delta oldsymbol{\Theta}$	=	uncertainty in model-multiplier parameter vector	
Δ_a	=	complex additive uncertainty	
Δ_i	=	complex input-multiplicative uncertainty	
Δ_o	=	complex output-multiplicative uncertainty	
Δ_r	=	real output-multiplicative uncertainty	
Δ_{ro}	=	mixed output-multiplicative uncertainty	
ϵ	=	minimum upper bound of ε	
arepsilon	=	frequency-dependent additive uncertainty	
Θ	=	model-multiplier parameter vector	
$\hat{oldsymbol{\Theta}}$	=	estimated model-multiplier parameter vector	
μ	=	structured singular value	
Φ	=	matrix of modal frequency responses	
ϕ	=	model frequency response vector	
•		* * *	

Introduction

A EROSERVOELASTIC systems comprise interactions of generally multi-input/multi-outputsampled-datacontrol feedback with actuation dynamics coupled with aeroelasticity. Highly augmented closed-loop flight-test data require extra care in distinguishing system component dynamics. Discrimination of source and re-

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sponse effects for proper understanding of issues in causality may be problematic. Flight-test verification of an aeroservoelastic model can also prove difficult in discerning the individual subsystem dynamics because of inaccessible parameters or inadequate sensing for system identification procedures. For instance, aerodynamic parameter identification algorithms for aeroelastic effects may encounter problems with flexibility, closed-loop coupling, and transonic nonlinearity.

Model verification over an extensive flight envelope presents more challenges. Test data acquisition is expensive, and so maneuvers are designed for maximum efficiency and data quality. A verification method is desired that accurately and efficiently includes identification of critical parameters, addresses mismodeling and unmodeled dynamics, deals with test condition and system variability, and derives data-consistent parametric and nonparametric uncertainty descriptions. Parametric uncertainty is generally caused by mismodeling of system properties, off-nominal test conditions, and model oversimplifications. Nonparametric uncertainty often relates to unmodeled dynamics and exogenous inputs and requires weaker assumptions on the identified system.

With such apparent complications in mind, this paper takes the approach that estimation of aeroservoelastic models must deal directly with uncertainty in model verification. Parameter identification will be applied for model updates from the test data while addressing mismodeling and unmodeled dynamics. Parametric and nonparametric uncertainty are incorporated to help minimize conservativeness and include both structured and unstructured uncertainty. Nonstatistical estimation approaches are preferred to avoid restrictive assumptions, minimize algorithmic complexity, and improve reliability in the form of error bounds. Most importantly, the effectiveness of model-based quantification of uncertainty bounds is appealing for robust control-oriented applications. ²⁻⁴

Set membership identification has been presented in a variety of contexts. Bounded error estimation,⁵ or bounded data uncertainty,⁶ characterizes feasible sets of parameters with uncertainty estimates consistent with the data, model structure, and prior information on uncertainty bounds. This last requirement can be in the form of unknown but bounded disturbances,^{7,8} constraints on the system impulse response and inputs,⁹ or assumptions on bounded data perturbations.¹⁰

Two general research directions of set membership estimation are 1) obtain the exact membership set and 2) compute a specific optimal estimate in the membership set. The former has suffered from

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computational complexity and conservatism.^{7,11} An optimal, robust minimax estimate approach is applied in this paper, but the a priori uncertainty bound is not required. Minimum upper error bounds are computed with the parameter estimates such that the feasible set is described as a function of the error bounds. Hence, computation of the minimum error bound results in a smallest nonempty feasible set.¹²

Transfer functions and modal parameter estimates derived from time-frequency representations have previously been applied to estimate state-space aeroservoelastic models. 13,14 Morlet wavelet filtering (see Refs. 15–17) is employed in this paper to update modal parameters as a first step in aeroservoelastic model identification and uncertainty estimation. Standard transfer functions are then employed in the estimation of input–output parameters with associated uncertainty using an optimal minimax procedure. Bounds derived from these estimates define parametric and nonparametric errors that relate to multiplicative and additive uncertainty structures, respectively, for mixed- μ (Refs. 18 and 19) robust stability analyses.

Uncertainty modeling for aeroservoelastic data analysis has not been addressed adequately in the literature. This paper addresses the problem by deriving models with nonconservative uncertainty descriptions consistent with the flight data in a robust control-oriented approach. F-18 Systems Research Aircraft (SRA)²⁰ data are used to demonstrate comparisons made between models derived from robust estimation methods using multiplicative and additive uncertainty structures.

Uncertainty Modeling

The aeroservoelastic open-loop plant model includes rigid-body and elastic modes, coupled high-order actuator dynamics, and control surface modal dynamics. ²¹ Including the aerodynamic lag states, the aeroservoelastic state equations take the following form:

$$\dot{x} = Ax + Bu, \qquad u = \delta_d$$

$$y = Cx + Du, \qquad x = \lceil \eta_\delta \eta_r \eta_e \dot{\eta}_r \dot{\eta}_e \eta_e \delta \rceil$$

consisting of input control surface commands δ_d , actuator states η_δ , rigid-body states η_r , flexible mode states η_e , aerodynamic lag states η_a , and control surface displacements δ . Aeroservoelastic plant P is, therefore, represented as the state-space operator. Associated with this time-domain representation is the transfer function P(s), a function of the complex Laplace variable s, such that y = P(s)u:

$$P(s) = D + C(sI - A)^{-1}B$$
 (1)

Controller K(s) is modeled similarly, but being a digital implementation of the aircraft control laws, it is modeled as a function of discrete complex variable z as $K(z = e^{sT})$, specified by the sampling time T and a zero-order sample hold at the input of the controller.

A robust characterization of the feedback model incorporates unstructured uncertainty to account for unmodeled dynamics and parameter variations. Unmodeled dynamics are represented with simple cone-bounded transfer functions at the input-output reference locations. Assume the model is suitably scaled with weightings W_1 and W_2 , so that the uncertainty can be represented by operator $W_1\Delta W_2$. With uncertainty incorporated into the proper loop reference locations, the robust stability condition is determined by analyzing unity-norm bounded perturbations, $\|\Delta\|_{\infty} < 1$, with the small gain theorem.¹⁹

Multiplicative uncertainty as shown in Fig. 1 is used to represent unmodeled dynamics and errors at the feedback output sensors $(w_o = \Delta_o z_o)$ and actuator input commands $(w_i = \Delta_i z_i)$. Each of Δ_o and Δ_i are diagonal complex perturbations of appropriate output or input dimensions. Performance specifications are in terms of sensor noise attenuation (output response to output commands) and

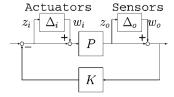


Fig. 1 Closed-loop model with input-output complex multiplicative uncertainty.

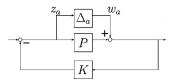


Fig. 2 Closed-loop model with additive aeroservoelastic plant uncertainty.

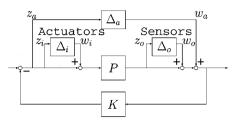


Fig. 3 $\,$ Closed-loop model with additive and complex multiplicative plant uncertainty.

actuator disturbance rejection (input response to input commands), respectively.

Multiplicative perturbation at the output results in perturbed model $\tilde{P}_o = (I + \Delta_o)P$, and at the input the perturbed model is $\tilde{P}_i = P(I + \Delta_i)$. Necessary and sufficient conditions for unstructured robust stability are then derived from tests on loop sensitivity functions. Unstructured robust stability tests representing these types of uncertainty are described from the complementary sensitivity matrix functions of complex P(s) and K(s):

$$\|\Delta_o\|_{\infty} < 1,$$
 $\|W_2 T_o W_1\|_{\infty} \le 1,$ $T_o = PK(I + PK)^{-1}$
 $\|\Delta_i\|_{\infty} < 1,$ $\|W_4 T_i W_3\|_{\infty} \le 1,$ $T_i = KP(I + KP)^{-1}$ (2)

Another uncertainty characterization of interest in this paper is the additive stable perturbation, Δ_a , for which the perturbed plant is $\tilde{P}_a = P + \Delta_a$, which is depicted in Fig. 2 as additive plant errors $(w_a = \Delta_a z_a)$. The corresponding control action robust stability test of input response to output disturbances from additive plant errors is imposed by the loop shape condition

$$\|W_6K(I+PK)^{-1}W_5\|_{\infty} \le 1$$
 (3)

The characterization that augments $(w_a = \Delta_a z_a)$ with the complex multiplicative uncertainty representation of unmodeled dynamics from Fig. 1 is shown in Fig. 3. Here, Δ_a is a full-block complex perturbation because frequency-dependent errors are allowed to enter any of the multi-input/multi-output loops in an arbitrary fashion.

The robust stability criteria of Eq. (2) vary with the uncertainty description. Uncertainty structure depends on the type of perturbation and how it connects to the nominal system. A mixed uncertainty structure consists of (real) parametric and (complex) unmodeled dynamic perturbations and cannot be treated adequately with a simple cone-bounded representation. The structured singular value μ is used to reduce conservatism for problems with structured specifications of uncertainty. Robust stability tests of Eq. (2) are stated in terms of an upper bound of μ_{Δ} (noting μ dependence on Δ) at each frequency ω (Ref. 18).

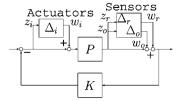
In this paper, the analysis setup of Fig. 1 is primarily used as a benchmark for aeroservoelastic stability analysis in the μ framework. In the output-multiplicative parameterization, $\tilde{P}_o = (I + \Delta_o)P$, for example, the relations

$$y = (I + \Delta_o) P u,$$
 $z_o = -PK(w_o + z_o) = -(I + PK)^{-1} PKw_o$

indicate that closed-loop stability is guaranteed if $(I + PK)^{-1}PK\Delta_o$ is less than unity. This suggests an interpretation as a multivariable transfer function gain in the sense that $\log \|T_o\Delta_o\| \le \log \|T_o\|$ because $\|\Delta_o\| < 1$.

To establish a common analysis consistent among the various uncertainty structures, while dealing with structured uncertainty in

Fig. 4 Input complex with output-mixed multiplicative uncertainty.



general, the robust stability criterion of Eq. (2) are replaced with

$$\mu_{\Delta_o}(W_2 T_o W_1) \le 1,$$
 $\|\Delta_o\|_{\infty} < 1$ $\mu_{\Delta_i}(W_4 T_i W_3) \le 1,$ $\|\Delta_i\|_{\infty} < 1$ (4)

and the control action stability criterion of Eq. (3) similarly becomes

$$\mu_{\Delta_a} [W_6 K (I + PK)^{-1} W_5] \le 1, \qquad \|\Delta_a\|_{\infty} < 1$$
 (5)

These conditions imply that for all perturbations matrices $\{\Delta_o, \Delta_i, \Delta_a\}$ with appropriate structure and satisfying the upper bound constraints, $\|\Delta_o\|_{\infty} < 1$, $\|\Delta_i\|_{\infty} < 1$, and $\|\Delta_a\|_{\infty} < 1$, respectively, the perturbed system is stable. All weights W_i are chosen to scale the auxiliary variables $\{w, z\}$ such that these upper bound constraints are valid. Also, there is a particular perturbation matrix not satisfying the contraints that causes instability, and this is found best from the computational lower bound. Is, 19 Therefore, the μ upper bound plot determines the size of perturbation for which the loop is robustly stable. Lower peaks imply more robust stability.

In the present application, real parameter uncertainty is represented with bounded real perturbations, $[\delta_1, \ldots, \delta_n]$, $|\delta_i|_{\infty} < 1$, in the aeroservoelastic plant output. A diagonal real-perturbation block is augmented to the complex output perturbation block Δ_a to get

$$\Delta_r = \begin{bmatrix} \delta_1 & & & \\ & \ddots & & \\ & & \delta_n \end{bmatrix}, \qquad \Delta_{ro} = \begin{bmatrix} \Delta_r & & \\ & \Delta_o \end{bmatrix}$$

$$w_{ro} = \Delta_{ro} z_{ro} \tag{6}$$

The multiplicative structure from Fig. 1 is modified with the additional $(w_r = \Delta_r z_r)$ as shown in Fig. 4. Therefore, Fig. 1 is actually contained in Fig. 4, but the additional real perturbation block at the output in Fig. 4 will account for individual modal contributions to the feedback response to the controller. A real- μ analysis augmented with a complex block is a mixed- μ problem. Complex blocks added to real perturbation problems have engineering relevance by accounting for phase uncertainty, besides guaranteeing continuity properties and assisting convergence. The complex blocks $\{\Delta_o, \Delta_i\}$ for the current problem are also motivated by uncertainty modeling of unmodeled dynamics as a function of the nominal complementary sensitivity transfer functions in Eq. (4) and, hence, are retained from Fig. 1.

Alternatively, the effect of real parametric uncertainty at the plant input or output will be shown to represent an additive uncertainty in the plant transfer function, as in Fig. 3. In the current analysis, the additive perturbation derived from the plant uncertainty $(w_a = \Delta_a z_a)$ will be compared with a consistent real perturbation analysis $(w_r = \Delta_r z_r)$ of the associated structured uncertainty of Fig. 4 for the F-18 SRA aeroservoelastic model.

Wavelet Filtering and Modal Identification

Time-frequency analysis provides a powerful tool for the analysis of nonstationary signals. ^{22–24} Signal structure is revealed by quantifying the time-frequency distribution of signal energy as a joint function of time and frequency. Energy density concentrations are revealed as specific areas in the time-frequency plane.

A novel multiresolution wavelet signal processing method is applied to time-frequency analysis of signals by decomposing data into cells with properties of scale and frequency concentrated in time. The cells consist of Gaussian-windowed sinusoidal basis functions, also known as Morlet wavelets, creating a multiscale decomposition in a filter bank structure. Competing requirements of time and frequency resolution, subject to the uncertainty principle, Si is

accomplished with a combination of dyadic multiscale decomposition, compact orthogonality, and harmonic wavelet properties.²⁵

Parameter estimates are derived from time-frequency representations using Morlet wavelet filtering (see Refs. 15 and 16). Morlet filtering is a signal analysistechnique that can use time-frequency input information for specification of energy concentrations but does not regard the system as an input-output state-space realization. Morlet wavelets consistute the basis for the energy-density distribution and, assuming the dominant sinusoidal characteristics in sensor responses are modal responses, are then used to estimate the modal parameters. The wavelet basis representation of the signal is, therefore, a projection subspace for extraction of modal dynamics.

As a first step in aeroservoelastic model identification and uncertainty estimation, the state-space transfer function plant description of Eq. (1) is transformed into real bidiagonal modal form with transformation matrix T:

$$P = \begin{bmatrix} A & B \\ \hline C & D \end{bmatrix}$$

$$=\begin{bmatrix} -(\zeta \omega_n)_1 & -\omega_{d1} & 0 & 0 & 0 \\ \omega_{d1} & -(\zeta \omega_n)_1 & 0 & 0 \\ 0 & \ddots & 0 & TB \\ 0 & 0 & -(\zeta \omega_n)_M & -\omega_{dM} \\ 0 & 0 & 0 & \omega_{dM} & -(\zeta \omega_n)_M \end{bmatrix}$$
(7)

These two-by-two blocks of complex conjugate roots represent M number of modes, where ζ is the modal damping ratio, ω_n the natural modal frequency, and $\omega_d = \omega_n \sqrt{(1-\zeta^2)}$ is the damped modal frequency for each mode (ignoring real roots for simplicity here). From this state-coordinate transformation the roots of structural modes are generally simple to discriminate from actuator, rigid-body, and aerodynamic lag states.

Modal parameters estimated from Morlet wavelet filtering of the open-loop aeroservoelastic plant transfer function are incorporated into the A-matrix modal form under the following guidelines:

- 1) Estimated modal frequency lies within specified range.
- 2) Modal parameter estimate confidence factor is sufficient.¹⁵
- 3) Search over ω_d for nearest estimate to A-matrix imaginary pairs.
- 4) A-matrix complex conjugate pair must correspond to a structural mode.
- 5) A-matrix modal pairs are replaced with estimates under criteria 1-4.

In this paper, parametric errors in modal frequency and damping estimation are not explicitly considered in the uncertainty description, as in previous studies. ^{16,21} Justification for this lies in choosing a high confidence factor for allowable estimates to help minimize estimation error. ¹⁵ Also, this error will be implicit in the uncertainty model development to be discussed.

Observability of modal dynamics obviously affects identifiability. All available control feedback responses of the aeroservoelastic plant are used to identify modal parameters from each control command maneuver. Numerous reponses become available for estimation of most modes to establish a high degree of confidence from at least one of the responses.

For the F-18 SRA maneuvers used in this research, Table 1 lists the available control commands used to generate the listed feedback signals for modal parameter estimation and subsequent transfer functionestimation for uncertainty analyses. Discrete multisine control commands are the inputs for the aeroservoelastic transfer functions. There are 25 analysis maneuvers in the matrix. Subsets of these correspond to predominantly symmetric or antisymmetric maneuvers, but, in reality, most modes are excited with either type of input.

Minimax Parameter and Uncertainty Estimation

Uncertainty estimation for aeroservoelastic systems depends on a variety of off-nominal factors to consider, such as fuel weight, flight condition, control gains, hinge moments, and other aerodynamic

Table 1 Feedback measurements and control commands used for the F-18 SRA aeroservoelastic plant transfer function and modal parameter estimates

Feedbacks	Control commands
Long	itudinal
Pitch rate	Symmetric stabilator
Normal acceleration	Symmetric aileron
Lateral-	directional
Roll rate	Differential aileron
Yaw rate	Differential stabilator
Lateral acceleration	Rudder

effects. The maneuvers used in this research from the F-18 SRA were predominantly flown for the purpose of model verification and update in support of the F-18 Active Aeroelastic Wing Program (AAW).²⁶ Flight conditions, therefore, lie entirely inside the F-18 AAW envelope, which is predominantly transonic and near transonic. Aerodynamic parameter estimation, loads analysis, and aeroservoelastic model verification were primary flight-test objectives for these maneuvers.

System identification with uncertainty modeling requires determination of reliable bounds with the nominal estimates. Interval and bounded-error estimation techniques have predefined bounds on the error from a priori knowledge of the system or sensors. This assumption is unacceptable, especially for analysis of aeroservoelastic flight data in the transonic regime because reliable bounds on the errors do not exist. Most important, popular methods for identification are not model based, and so the errors are not relevant to the model, but only depend on the data and estimation process itself.

Aeroservoelastic models are often of high order, significant dynamic range, and contain lightly damped modes. Any methods that depend on simultaneous parameter and model order estimation will have complexity problems attempting to discern order from uncertainty. Even for fixed-order estimation, identification of structural parameters with input–output (C and B) parameters is susceptible to nonuniqueness, convergence, and bias problems unless severe assumptions are imposed. Again, these procedures are often based on statistical arguments and often are not model based. Stability prediction based on errors between models and aircraft requires uncertainty relative to a model.

Robust Minimax Parameter Estimation

Robust minimax estimation not only provides the minimum upper bound on the error, but also provides a parameter set compatible with any error upper bound. The problem is stated in terms of error $\varepsilon(k,\Theta)$, its bound ϵ , and parameter vector Θ , given N measured frequency response data y(k), and model frequency response vector $\phi(k)$:

$$\epsilon \equiv \|\varepsilon\|_{\infty} = \max_{k} |\varepsilon|,$$
 $\varepsilon(k, \Theta) = y(k) - \phi(k)\Theta$
$$k = 1, \dots, N \quad (8)$$

Because the minimum upper error bound over frequency ϵ is assumed unknown, it is estimated from the minimax procedure. The minimum value of ϵ is desired that is consistent with the flight data frequency responses. A solution set is given by

$$\left\{ \hat{\mathbf{\Theta}} | \hat{\mathbf{\Theta}} = \arg\min_{\Theta} \max_{k} |\phi(k)\mathbf{\Theta} - y(k)| \right\}$$
 (9)

which can be transformed to a differentiable linear programming problem using additional variable *x* subject to constraints

$$\hat{\mathbf{\Theta}} = \arg\min_{\mathbf{\Theta}}(x), \qquad |\phi(k)\mathbf{\Theta} - y(k)| \le x, \qquad k = 1, \dots, N$$

It can then be shown that the set of all (x, Θ) consistent with the constraints is a convex unbounded polyhedron, or for any $x \ge \epsilon$ (Ref. 12)

$$\mathbf{\Theta} = \hat{\mathbf{\Theta}} + \Delta \mathbf{\Theta} = \hat{\mathbf{\Theta}} + \Delta(x) \tag{10}$$

The feasible polytope for (x, Θ) , thus, contains the exact description of the solution set of $\hat{\Theta}$, including a range of feasible parameters contained in a positive interval $\Delta\Theta$ that satisfies Eq. (9) for any $x \ge \epsilon$. Therefore, the most important property of this approach for uncertainty estimation is that a minimum upper bound on the frequency response error ϵ is found that is compatible with parametric errors $\Delta\Theta$, so that these parametric errors are derived from the value of the nonparametric error cost function $|\epsilon(k, \Theta)|$.

A seemingly attractive analytic center approach to bounded-error estimation was recently proposed^{7,28} to minimize logarithmic average output error

$$\hat{\mathbf{\Theta}} = \arg\min_{\Theta} \sum_{k} \log |\phi(k)\mathbf{\Theta} - \mathbf{y}(k)|$$

This estimator has nice properties in terms of output error minimization, robustness to outliers, and online sequential implementations. However, these properties and error bounds depend on a priori knowledge of noise bounds. This estimator is not chosen because of the necessary noise assumptions and absence of guaranteed parametric error bounds.

Model Updates and Output-Uncertainty Estimation

Now the procedure for flight data analysis is described in the robust minimax estimation framework. First, the A matrix of Eq. (7) is updated to get estimate \hat{A} , with wavelet modal filtering. Then the column of the $\{B,D\}$ matrices in Eq. (7) corresponding to a particular control command input is appropriately scaled by matching the norms of the model and estimated data transfer functions from control command to feedback sensor. These estimates are denoted as $\{\hat{B},\hat{D}\}$. Each element of the model C matrix corresponds to a modal contribution to the feedback response. Elements of the appropriate row of the model C matrix corresponding to the feedback response are then chosen for optimization only if they correspond to modal responses within a specified frequency range. Vector \mathbf{y} is the flight data response from a feedback sensor, being a sum of the aircraft modal responses.

Each row of the model C matrix is expanded in a matrix diagonal

$$[c_{1}\cdots c_{n}] \rightarrow \begin{bmatrix} c_{1} & 0 & 0 \\ 0 & \ddots & 0 \\ 0 & 0 & c_{n} \end{bmatrix} \Longrightarrow \hat{P} = \begin{bmatrix} \hat{A} & \hat{B} \\ c_{1} & 0 & 0 \\ 0 & \ddots & 0 \\ 0 & 0 & c_{n} \end{bmatrix} \tilde{\hat{D}}$$

where $\tilde{\hat{D}}$ is resized from elements of \hat{D} corresponding to diagonalization of a row of C. Arrange the updated model frequency response matrix $\hat{P}(i\omega_k)$ to form the n columns (for n states) of matrix Φ , where each column corresponds to a modal frequency response contribution to the total feedback sensor response such that

$$\Phi(k, l) = \hat{P}(i\omega_k, l), \qquad l = 1, ..., n, \quad k = 1, ..., N$$
 (11)

Parameter vector $\boldsymbol{\Theta}$ is the multiplier (nominally $\boldsymbol{\Theta} = [1 \cdots 1]^T$) on the model response matrix $\boldsymbol{\Phi}$ to match the flight data response. Absolute transfer function error $\boldsymbol{\epsilon}$ is the maximum of the frequency-dependent error vector $\hat{\boldsymbol{\epsilon}}$ expressed as [compare to Eq. (8)]

$$\epsilon \equiv \|\hat{\varepsilon}\|_{\infty} = |\hat{\varepsilon}|, \qquad \hat{\varepsilon} = y - \Phi \hat{\Theta} = y - \hat{y}$$
 (12)

In light of the robust properties of logarithmic error criteria^{28–30} and the well-known property of Chebyshev estimators being optimal in terms of worst-case parameter error, the objective function is chosen as a log-type Chebyshev estimator:

$$\hat{\mathbf{\Theta}} = \arg\min_{\mathbf{\Theta}} \max_{\mathbf{\omega}} \log |\mathbf{\Phi}\mathbf{\Theta} - \mathbf{y}| \tag{13}$$

There is strong justification for this performance criterion. It has been shown that logarithmic error minimization is superior to

absolute error minimization for parametric transfer function estimation, especially for structural systems and high-order problems.³⁰ Properties from Eq. (9) still hold in a log sense because the log function and its inverse (exponential) are monotonic smooth functions of the data. Furthermore, log-based frequency-response estimation demonstrates superior performance in closed-loop systems, with mutually correlated signals in feedback, and is robust to noise characteristics and outliers without relying on statistical assumptions.²⁹

Parametric uncertainty is multiplicative in the real C-matrix values because the identified parameter vector $\hat{\Theta}$ is a multiplier to the model. This is the optimal nominal estimate. It was also shown that from the identification procedures of Eq. (9) or Eq. (13), a real positive uncertainty interval $\Delta \Theta$ to this nominal part results from Eq. (10). Decomposing the identification problem into these nominal and uncertain parts, an estimate of the real parameter uncertainty becomes [see Eq. (12)]

$$\|\hat{\mathbf{y}}\| + \epsilon \equiv \|\Phi\hat{\boldsymbol{\Theta}}\| + \|\Phi \cdot \Delta\boldsymbol{\Theta}\| \Longleftrightarrow \epsilon = \|\Phi \cdot \Delta\boldsymbol{\Theta}\| \le \|\Phi\|\|\Delta\boldsymbol{\Theta}\|$$

$$\Longrightarrow \Delta \Theta \approx \epsilon \cdot / \sum_{k=1}^{N} |\Phi_{k,l}|, \qquad l = 1, \dots, n \quad (14)$$

where element-by-element division $(\epsilon./)$ is implied. In the expression $\Phi \cdot \Delta \Theta$, each uncertain element of $\Delta \Theta$ is weighted by a column sum of Φ . Each element of the positive uncertain vector $\Delta \Theta$, therefore, results from a division by

$$\sum_k |\Phi_{k,l}|$$

a weighting by an inverse absolute column sum (vector 1-norm). This is a decomposition of the absolute total error over frequency ϵ into individual modal contributions in the elements of $\Delta \Theta$. Elements of $\Delta \Theta$ with denominator sums less than a small threshold retain nominal $\hat{\Theta}$ values $[\Delta \Theta(l)=0]$ to avoid numerical problems, but also denote relatively insignificant modal responses [column sums of $|\Phi|$, see Eq. (11)] with very low observability.

Note that $\Delta\Theta = \Delta(\epsilon)$ from the linear programming solution [compare Eqs. (10) and (14)], so that the parametric error is a byproduct of the nonparametric frequency-dependent global error $\hat{\epsilon}$. Also note that the minimum upper bound error ϵ is a constant additive perturbation bound over frequency from the relations in Eqs. (8) and (12). This corresponds to the uncertainty that is added to the nominal estimate, $\hat{y} = \Phi\hat{\Theta}$, to account for the flight data transfer functiony from Eq. (8). In either case, the nonparametric frequency-dependent error $\hat{\epsilon}$, or its more conservative bound ϵ , characterizes the additive perturbation error at each frequency in Figs. 2 and 3:

$$w_a = \Delta_a z_a = \Delta_a |\hat{\varepsilon}| \le \Delta_a \epsilon \tag{15}$$

Alternatively, the real-parametric error $\Delta \Theta$ naturally represents real uncertainty from Eq. (6) and Fig. 4 as a multiplier on the nominal C matrix. For each output row, using elementwise multiplications denoted by (.*), the uncertainty is described by

$$C_{\Delta} \equiv \hat{C} \pm \Delta C = C. * \hat{\mathbf{\Theta}} \pm \Delta C, \qquad |\Delta C| \equiv |C. * \Delta \mathbf{\Theta}|$$
$$\|\Delta C\| \le \|C. * [\Delta_r \Delta \mathbf{\Theta}]\| \tag{16}$$

where operator Δ_r is the real perturbation defined in Eq. (6), $w_r = \Delta_r z_r = \Delta_r \Delta \Theta$, $\|\Delta_r\| < 1$, and, therefore, $\|\Delta_{ro}\| < 1$ because $\|\Delta_o\| < 1$. Note that real-parametric variation ΔC varies equally positively or negatively according to $\|\Delta_r\| < 1$, whereas $\Delta \Theta$ is a positive range of variation.

Note here that $\Delta\Theta$, therefore, serves both as a contribution to the additive error bound [from Eqs. (14) and (15), allowed to be complex in general to account for phase variations] and as a multiplicative-output real-parametric error in Eq. (16), and so it is integral to both types of uncertainty descriptions. Despite this relationship, the manner in which the uncertainty is modeled (complex additive vs real multiplicative) determines its structure and will affect the robust stability analysis. Comparisons of both manifestations will be demonstrated next.

Aeroservoelastic Flight Data Analysis

Three types of uncertainty structures have been discussed as being relevant to aeroservoelastic uncertainty modeling. They are compared in this section: 1) complex-multiplicative loop uncertainty [baseline structure, Fig. 1, with criteria in Eq. (4)], 2) additive with complex-multiplicativeloop uncertainty [Fig. 3 with Δ_a in Eq. (15)], and 3) mixed real parametric with complex-multiplicative loop uncertainty [Fig. 4 with Eq. (16)].

Complex-multiplicative loop uncertainty at the input and output in Fig. 1 is the baseline analysis. This baseline structure is augmented with additional perturbation blocks in the additive and mixed analyses. All weights W_i are unity by nature of the perturbation descriptions, maintaining that $\|\Delta\| < 1$ for all perturbations, $w = \Delta z$. All uncertainty analyses operate on models that are updated with the wavelet modal parameter estimates unless stated otherwise. Results are for the F-18 SRA flight condition of Mach 0.9, altitude 5000 ft using all three available inputs denoted in Table 1 for lateral-directional maneuvers.

Baseline complex-multiplicative complementary sensitivity μ -analysis results, using a discrete implementation of controller K as mentioned after Eq. (1), are compared with additive loop uncertainty at the output (solid) and input (dashed) in Fig. 5 for the F-18 SRA lateral-directional flight condition. Comparisons between upper-bound additive (ϵ ; Fig. 5a) and varying additive ($\hat{\epsilon}$; Fig. 5b) results are presented. Note that the additive results must be at least as large in magnitude as the baseline because additive includes complex-multiplicative uncertainty (Fig. 3) and, therefore, includes Δ_a as an additional perturbation to the baseline structure. These plots demonstrate that the estimated minimax error upper bound can be somewhat conservative compared to the frequencydependent varying bound, but not too unreasonable compared to the varying error result. Error bounds from the additive minimax analysis of the flight data at this condition are evident as being up to 15 dB larger than the baseline result.

Additive and mixed uncertainty are compared with complementary sensitivity complex-multiplicative μ -analysis results in Fig. 6, again for the F-18 SRA lateral-directional flight condition. Also shown in these plots are the nominal complex-multiplicative results computed from aeroservoelastic models before any updates from flight-derived parameter estimates were incorporated.

In Fig. 6, the dominant modes are antisymmetric fuselage first bending (near 8 Hz) and antisymmetric wing first bending (near 9 Hz). Differences in modal pole-zero relationships and modal frequency shifts between the nominal and updated (using $\{\hat{A}, \hat{B}, \hat{C}, \hat{D}\}$) complex-multiplicative results demonstrate the significant effect of parametric updates on the nominal model. As expected, both mixed and additive results are at least as large in magnitude as the baseline. Although the output-mixed- μ magnitude is the largest (solid thick line in Fig. 6a) and much larger than the output-additive result (solid thick line in Fig. 6b), the additive results are

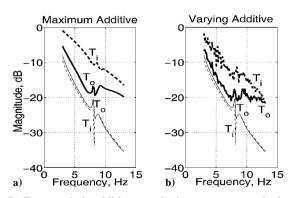


Fig. 5 For μ -analysis: additive perturbations, output-perturbationanalysis (T_o) , input-perturbation analysis (T_i) : a) complex-multiplicative uncertainty (—; structure in Fig. 1) compared to upper-bound ϵ -additive μ -analysis with complex-multiplicative uncertainty (—; structure in Fig. 3) and b) complex-multiplicative uncertainty (—; structure in Fig. 1) compared to varying $\hat{\epsilon}$ -additive μ -analysis with complex-multiplicative uncertainty (—; structure in Fig. 3).

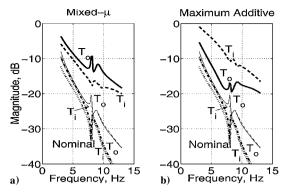


Fig. 6 For μ -analysis: mixed and additive perturbations, output-perturbation analysis (T_o) , input-perturbation analysis (T_i) : a) complex-multiplicative uncertainty (—; structure in Fig. 1) compared to mixed real parametric with complex-multiplicative uncertainty (—; structure in Fig. 3) and nominal complex-multiplicative uncertainty and b) complex-multiplicative uncertainty (—; structure in Fig. 1) compared to upper-bound ϵ -additive μ -analysis with complex-multiplicative uncertainty (—; structure in Fig. 3) and nominal complex-multiplicative uncertainty.

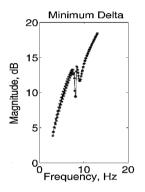


Fig. 7a Mixed uncertainty minimum perturbation to instability.

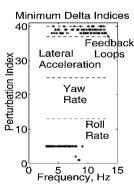


Fig. 7b Index of perturbation corresponding to output mixed- μ analysis from the left-hand plot in Fig. 6.

reversed in that the input-additive magnitude (dashed thick line in Fig. 6b) dominates with values above all of the other results. Despite the relationship between the mixed and additive uncertainties $[\Delta \Theta \Longleftrightarrow \epsilon; \sec \text{Eq. } (14)]$ the connection is disguised by the manner in which the uniform additive uncertainty, ϵ from Eq. (15), is incorporated in the structure of Fig. 3 compared to the mixed augmentation of $\Delta \Theta$ in Eq. (16) and Fig. 4.

Analysis of numerous flight conditions does not reveal any consistent trends between the mixed and additive results. Quality of models and flight data, as well as loop structure (longitudinal or lateral-directional), are primary considerations. For example, transonic conditions often exhibit larger uncertainty because of the model sensitivity²¹ and the relatively poor quality of flight data from exogenous inputs due to aerodynamic effects. Another source of significant uncertainty can be dynamic modal cross-axis coupling, especially with modes closely spaced in frequency, between symmetric and antisymmetric modes. Aeroelastic and aeroservoelastic models are often developed with a preferred axis orientation, symmetric or antisymmetric. Cross-axis dynamics will necessarily be revealed as unmodeled dynamics in such a situation.

Lower bounds of μ provide the perturbation Δ , which causes instability¹⁸ with respect to its structure. [See comments following Eq. (4).] In Fig. 7a, lower bounds are calculated and plotted with line-connected asterisk symbols. Lower values of the minimum perturbation denote the more sensitive frequencies. In Fig. 7b, corresponding indices of the real parameters ($w_r = \Delta_r z_r$, indices 1–37) and complex uncertainty control feedback loops (three output loops, $w_a = \Delta_a z_a$ with indices 38-40) are marked, also plotted with individual asterisk symbols. This information shows which parametric uncertainties (indices 1-37) or complex-multiplicative loop uncertainties (indices 38-40) in each feedback, at specific frequencies, result in the minimum perturbation-to-instability condition of Fig. 7a. For example, complex-multiplicative uncertainty in all output loops (indices 38-40) is very significant, as seen from a congestion of values at the top of Fig. 7b (three rows corresponding to indices 38-40 at multiple frequencies). Real parameters $\Delta \Theta$ contributing to roll rate are the other perturbations important in the instability mechanism, as seen from the row and scatter of points plotted at the bottom. There are no critical real perturbations in the yaw rate (indices 14-25) or lateral acceleration (indices 26-37) loops.

Conclusions

Aeroservoelatic model identification with uncertainty is addressed to present a robust data-oriented procedure for model development. Surface command inputs and control system feedbacks are used as signals in a wavelet-based modal estimation procedure for modal parameter updates. Transfer functions are incorporated in a robust minimax estimation scheme to identify input-output parameters and structure derror bounds consistent with the data. Uncertainty estimates derived from the data in this manner provide appropriate and relevant representations for robust stability analysis useful for model validation and control system design. This procedure is an automated, efficient, and reliable approach for analysis of numerous flight data sets for robust stability and model development.

References

¹Britt, R. T., Jacobsen, S. B., and Arthurs, T. D., "Aeroservoelastic Analysis of the B-2 Bomber," *Journal of Aircraft*, Vol. 37, No. 5, 2000, pp. 745–752.

²Andersson, L., and Rantzer, A., "Frequency-Dependent Error Bounds for Uncertain Linear Models," *IEEE Transactions on Automatic Control*, Vol. 44, No. 11, 1999, pp. 2094–2098.

³Campbell, M. E., and Grocott, S. C. O., "Parametric Uncertainty Model for Control Design and Analysis," *IEEE Transactions on Control Systems Technology*, Vol. 7, No. 1, 1999, pp. 85–96.

⁴Smith, R. S., and Dahleh, M. (eds.), *The Modeling of Uncertainty in Control Systems*, Springer-Verlag, London, 1994, pp. 1–75.

⁵Milanese, M., Horton, J., Piet-Lahanier, H., and Walter, É. (eds.), *Bounding Approaches to System Identification*, Plenum, New York, 1996, pp. 183–197.

⁶Sayed, A. H., and Chandrasekaran, S., "Parameter Estimation with Multiple Sources and Levels of Uncertainties," *IEEE Transactions on Signal Processing*, Vol. 48, No. 1, 2000, pp. 680–692.

⁷Bai, E.-W., Yue, Y., and Tempo, R., "Bounded Error Parameter Estimation: A Sequential Analytic Center Approach," *IEEE Transactions on Automatic Control*, Vol. 44, No. 6, 1999, pp. 1107–1117.

⁸Tan, G., Wen, C., and Soh, Y. C., "Identification for Systems with Bounded Noise," *IEEE Transactions on Automatic Control*, Vol. 42, No. 7, 1997, pp. 996–1001.

⁹Feng, X., Lin, C.-F., and Coleman, N. P., "Frequency-Domain Recursive Robust Identification," *Journal of Guidance, Control, and Dynamics*, Vol. 23, No. 5, 2000, pp. 908–910.

¹⁰Sayed, A. H., and Nascimento, V. H., "Design Criteria for Uncertain Models with Structured and Unstructured Uncertainties," *Robustness in Identification and Control*, edited by A. Garulli, A. Tesi, and A. Vicino, Vol. 245, Lecture Notes in Control and Information Sciences, Springer-Verlag, London, 1999, pp. 159–173.

¹¹Rotstein, H., Galperin, N., and Gutman, P.-O., "Set Membership Approach for Reducing Value Sets in the Frequency Domain," *IEEE Transactions on Automatic Control*, Vol. 43, No. 9, 1998, pp. 1346–1350.

¹²Walter, É., and Piet-Lahanier, H., "Recursive Robust Minimax Estimation," *Bounding Approaches to System Identification*, edited by M. Milanese, J. Horton, H. Piet-Lahanier, and É. Walter, Plenum, New York, 1996, pp. 183–197.

¹³Brenner, M. J., and Feron, E., "Wavelet Analysis of F/A-18 Aeroelastic and Aeroservoelastic Flight Test Data," AIAA Paper 97-1216, April 1997; also NASA TM-4793, 1997.

¹⁴Feron, E., Brenner, M., Paduano, J., and Turevskiy, A., "Time-Frequency Analysis for Transfer Function Estimation and Application to Flutter Clearance," Journal of Guidance, Control, and Dynamics, Vol. 21, No. 3, 1998, pp. 375-382.

¹⁵Brenner, M., "Nonstationary Dynamics Data Analysis with Wavelet-SVD Filtering," AIAA Paper 2001-1586, April 2001; also NASA TM-2001-

¹⁶Brenner, M., and Lind, R., "Wavelet-Processed Flight Data for Robust Aeroservoelastic Stability Margins," Journal of Guidance, Control, and Dynamics, Vol. 21, No. 6, 1998, pp. 823-829.

¹⁷Ruzzene, M., Fasana, A., Garibaldi, L., and Piombo, B., "Natural Frequencies and Dampings Identification Using Wavelet Transform: Application to Real Data," Mechanical Systems and Signal Processing, Vol. 11, No. 2, 1997, pp. 207-218.

¹⁸Balas, G. J., Doyle, J. C., Glover, K., Packard, A., and Smith, R., μ-Analysis and Synthesis Toolbox, MUSYN, Inc., Minneapolis, MN, and The Mathworks, Inc., Natick, MA, 1998, Chap. 4, pp. 4-1-4-86.

¹⁹Zhou, K., and Doyle, J. C., Essentials of Robust Control, Prentice-Hall, Upper Saddle River, NJ, 1998, Chaps. 8-10, pp. 129-220.

²⁰Sitz, J. R., F-18 Systems Aircraft Research Facility, NASA TM-4433, Dec. 1992.

²¹Lind, R., and Brenner, M., Robust Aeroservoelastic Stability Analysis: Flight Test Applications, Advances in Industrial Control Series, Springer-Verlag, London, 1999.

²²Hubbard, B. B., *The World According to Wavelets*, A. K. Peters, Wellesley, MA, 1998, Chap. 5, pp. 76–89.

²³Mallat, S., A Wavelet Tour of Signal Processing, Academic Press, New

York, 1999, Chap. 4, pp. 64-126.

²⁴ Vetterli, M., and Kovačević, J., Wavelets and Subband Coding, Prentice-Hall, Upper Saddle River, NJ, 1995, Chap. 6, pp. 333-368.

²⁵Le, D. K., "Application of Sampling Theorems in Wavelet Spaces to Multiresolution Visualization and Data Segmentation," SPIE Proceedings of Wavelet Applications in Signal and Image Processing III, Vol. 2569, Society of Photo-Optical Instrumentation Engineers, Bellingham, WA, 1995,

pp. 220–233.

²⁶Pendleton, E., Bessette, D., Field, P., Miller, G., and Griffin, K., "Active Aeroelastic Wing Flight Research Program: Technical Program and Model Analytical Development," Journal of Aircraft, Vol. 37, No. 4, 2000, pp. 554-

²⁷Zhang, Y. M., Kovacevic, R., and Yao, Y. X., "Identification of Multivariable Interval Dynamic Model from Measurements in Frequency Domain," Proceedings of the 1997 American Control Conference, Inst. of Electrical and Electronics Engineers, Albuquerque, NM, 1997, pp. 1939-

²⁸Bai, E.-W., and Yue, Y., "Constrained Logarithmic Least Squares in Parameter Estimation," *IEEE Transactions on Automatic Control*, Vol. 44, No. 1, 1999, pp. 184-186.

²⁹Guillaume, P., Pintelon, R., and Schoukens, J., "Robust Parametric Transfer Function Estimation Using Complex Logarithmic Frequency Response Data," IEEE Transactions on Automatic Control, Vol. 40, No. 7, 1995, pp. 1180–1190.

³⁰Sidman, M. D., DeAngelis, F. E., and Verghese, G. C., "Parametric Sys-

tem Identification on Logarithmic Frequency Response Data," IEEE Transactions on Automatic Control, Vol. 36, No. 9, 1991, pp. 1065-1070.